

Digital Frequency Comparator

This application note illustrates how a hierarchical all-digital design with two implementation views, can be defined in OrCAD Capture, and subsequently simulated in PSpice. The example circuit is a basic frequency-comparator (see Figure 1). All parts used in the schematic are provided in the standard Symbol and Model Libraries. One implementation is chosen for PSpice simulation to demonstrate the circuit's behavior.

Introduction

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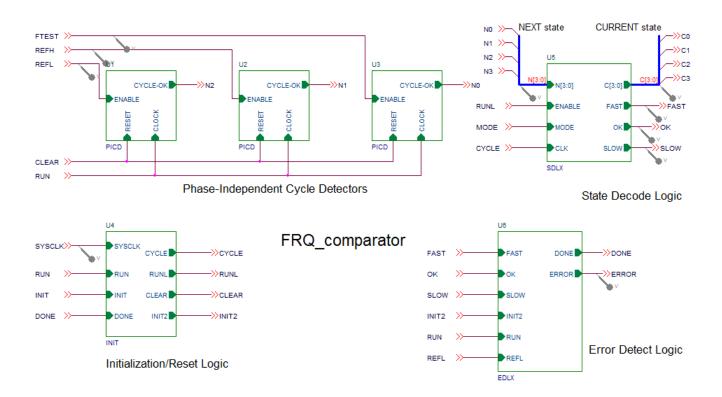


Figure 1: Top-level schematic for the frequency-comparator circuit

The frequency-comparator circuit accepts two reference frequency inputs, and a test frequency input which is compared to the references. After initialization and start-up, the circuit produces fast, slow, OK, and error indications. Operation is continuous as long as both of the reference signals are applied. Initialization is accomplished by applying a low pulse to the INIT input, having a minimum width of 40 nsec. At least 40 nsec after the negative-going edge of the INIT input, circuit operation commences upon applying a negative-going edge to the RUN input. Outputs of the circuit–SLOW, FAST, OK, and ERROR—are pulses indicating the result of comparing the test frequency signal, FTEST, to the low and high frequency reference signals, REFL and REFH, respectively. The ERROR pulse is generated if more than 7 complete periods of the REFL signal are observed with no activity on the FTEST input during that time.

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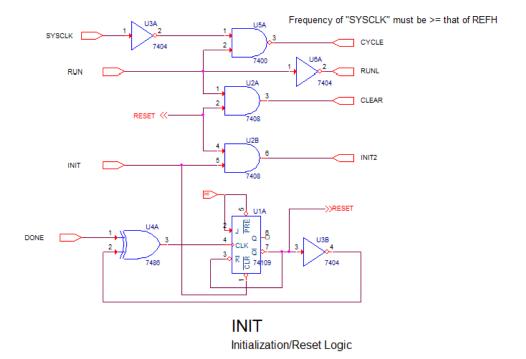


Figure 2: INIT block implementation

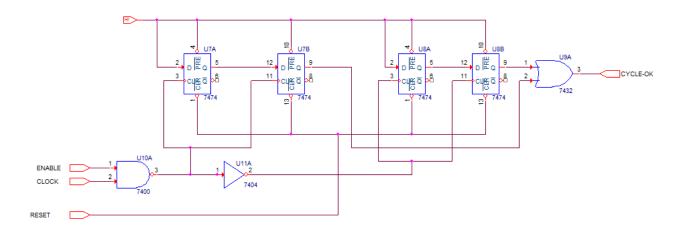


Figure 3: PICD block implementation

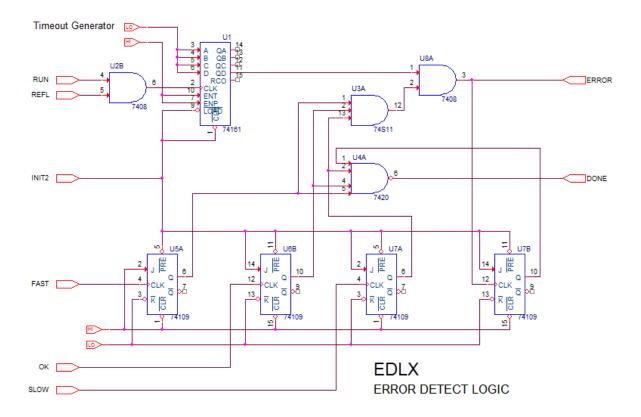


Figure 4: Error Detect Logic

Implementation

The frequency-comparator circuit is designed in OrCAD Capture using hierarchical blocks for the initializer (INIT block), cycledetectors (PICD blocks), state-decoder (SDL block), and errordetector (EDL block). The design has two alternative implementations: a gate-level implementation using off-theshelf 74xx parts (see Figure 5), and a functionally equivalent implementation using a mixture of 74xx parts and a commonly available Programmable Array Logic (PAL) device, PAL20RP4B (see Figure 6). Both implementations use the digital stimulus include file, *Freq_comparator.stm*, providing definitions for the INIT, RUN, MODE, REFH, REFH, FTEST, and SYSCLK input signals. The design alternatives are implemented as two views of the SDL block, with the DEFAULT view being the gate-level implementation, and the PAL-IMPL view being the PAL implementation. For the PAL-IMPL view, the data required to program the PAL20RP4B device is supplied in a JEDEC file, *FRQCHK.JED*, generated using OrCAD/PLD.

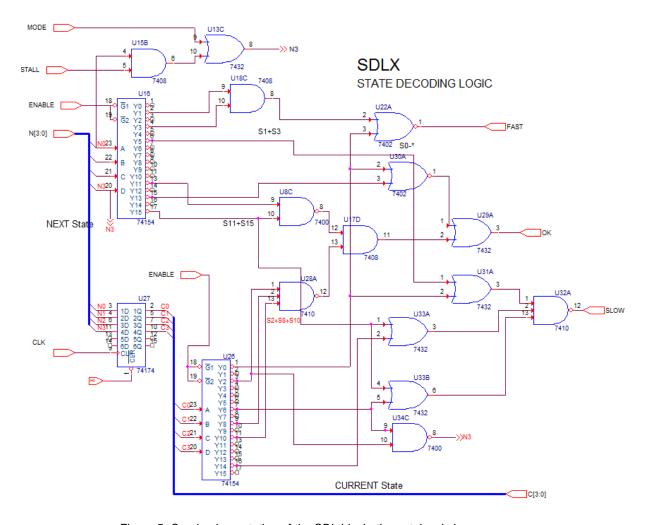


Figure 5: One implementation of the SDL block-the gatelevel view

SDLPX STATE DECODING LOGIC - PAL IMPLEMENTATION

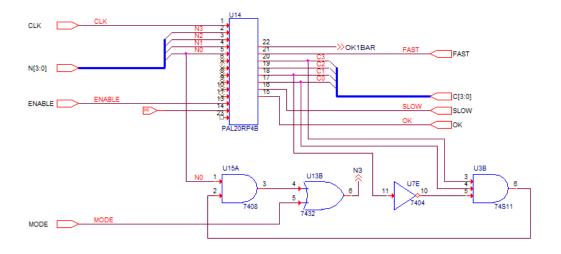


Figure 6: Another implementation of the SDL block-the PAL view

Operation

The three frequency inputs—REFL, REFH, FTEST—each drive a separate instance of a cycle-detector circuit (PICD blocks). Each cycle-detector is made up of two pairs of D-type flip-flops and a few basic gates. After having been *reset* and *enabled*, the cycle-detectors output a HI level as soon as two similar edges (e.g., falling) have been applied. This indicates that one complete period of the input signal has been observed. The circuit implements a simple finite-state machine (see Figure 7) that recognizes the order in which the individual frequency inputs make complete cycles.

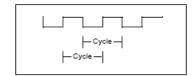
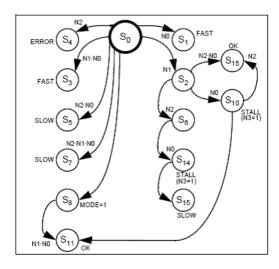


Figure 7: A Complete Cycle

For example, suppose that the REFH signal period is observed first (generating N1), followed by the REFL signal period (generating N2), then the FTEST period (generating N0). This indicates that the FTEST frequency is too low and that the SLOW output should be pulsed. But if the FTEST period is observed before the REFL cycle, an OK pulse is produced. The state machine *current state* simply represents the order of activity that has been observed since the last *initialization* or *reset*, which occurs every time any kind of output pulse is generated. The cycle-detectors monitor the input activity and produce the *next state* value (N3, N2, N1, N0), which is fed to the state-decoder (SDL block). At a rate determined by the system clock, SYSCLK, this *next state* becomes the *current state*; the 74154 4/16 decoders in the gate-level view of the state-decoder, continually provide unary logic indications of the next/current *transitions* (since *next state* values are not synchronized to SYSCLK). The random combinational logic in this same view recognizes the specific transitions that comprise the conditions of interest, i.e., FAST, SLOW, and OK, as per the state-transition diagram. (In the PAL view, the PAL20RP4B device replaces all of the decoding logic as well as the 4-bit register representing the *current state* value. The alternative implementations are functionally identical.) Note that the output indicators are not *static* state assignments; they are derived from selected state transitions. Thus, S14 þ S15 recognizes a SLOW condition, while S10 þ S15 signifies an OK condition.

The error-detector logic (EDL block) waits for the TIMEOUT signal output by the timeout generator. The timeout generator is simply a counter whose Q3 output indicates that the 8th rising edge of the low frequency reference, REFL, has occurred. If none of the normal output indicators (SLOW, FAST, or OK) have occurred before TIMEOUT, the ERROR output is asserted. The error-detector also asserts its DONE output whenever any of FAST, SLOW, OK, or ERROR have occurred.



The initialization/reset logic (INIT block) performs two functions. One distributes the effects of the INIT and RUN inputs, as defined in the stimulus include file, *Freq_comparator.stm*. The other uses the DONE signal from the error-detector to generate a RESET pulse; this has the same effect as the external RUN pulse—to restore the state machine to its starting state (0) as well as reset the cycle-detectors, timeout generator, and flip-flops in the error-detector. Normal operation then resumes.

JEDEC file containing the programming for the PAL20RP4B

```
OrCAD PLD 386
Type: PAL20RP4B
QP24* QF2568* QV1024*
F0*
L2560 11 11 11 11 *
C4B0E*
CCF0
```

Stimuli for the INIT, RUN, MODE, REFL, REFH, FTEST, and SYSCLK inputs

```
* "frqchk.stm" stimulus file

* uh1 stim (4,1111) $g_dpwr $g_dgnd

+ INIT RUN MODE REFL

+ IO_STM IO_LEVEL=0

+ 0s 1100

+ 2055ns 0100

+ 2135ns 1000

+ 2215ns 1000

+ 2255ns 1100
```

- + 5us 1101
- + label=loop1
- + +10us 1100
- + +10us 1101
- + +10us goto loop1 -1 times

uh2 stim (1,1) \$g_dpwr \$g_dgnd

- + REFH
- + IO STM IO LEVEL=0
- + 0s 0
- + +3us 1
- + label=loop1
- + +5us 0
- + +5us 1
- + +5us goto loop1 -1 times

uh3 stim (1,1) \$g dpwr \$g dgnd

- + FTEST
- + IO_STM IO_LEVEL=0
- + 0s 0
- + label=loop1
- + +20us 1
- + +20us 0
- + +20us goto loop1 5 times
- + + 0s 1
- + label=loop2
- + +3us 0
- + +3us 1
- + +3us goto loop2 20 times
- + +0s 1
- + label=loop3
- + +6us 0
- + +6us 1
- + +6us goto loop3 10 times

uh5 stim (1,1) \$g_dpwr \$g_dgnd

- + SYSCLK
- + IO STM IO LEVEL=0
- + 0s 0
- + +2us 0
- + label=loop1
- + +800ns 1
- + +800ns 0
- + +800ns goto loop1 -1 times

PSpice Simulation—PAL View

The transient analysis is defined with: Print Step = 1us and Final Time = 1ms. All flip-flops must be initialized in the 0 state (rather than the default X state). This allows the simulator to properly initialize the circuit by forcing the reset logic to a deterministic state (non X; the hardware implementation would eventually syncronize itself to the input stimuli and operate correctly). In the top-level schematic, the SDL block is the only block with more than one view. Without further setup, OrCAD Capture will generate the PSpice netlist using the DEFAULT gate-level view for SDL. After running the simulation by PSpice, the state-machine operation is viewed in Probe by placing markers on the appropriate wires and buses, or by typing the signal names in the Probe dialog under the Trace/Add command as follows:

SYSCLK, REFH, REFL, FTEST FAST, SLOW, OK, ERROR {N3, N2, N1, N0}; NEXT {C3, C2, C1, C0}; CURRENT

Figure 9 demonstrates the correct response of the circuit to the digital stimulus at FTEST.

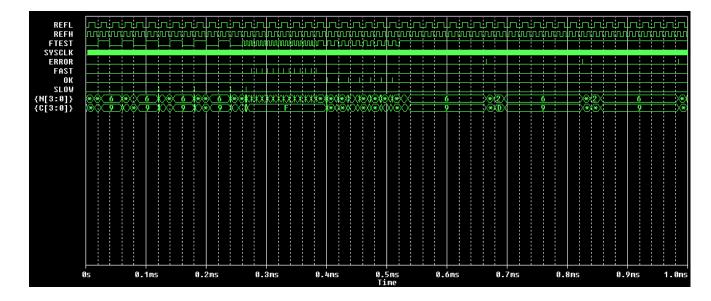


Figure 9: Frequency-comparator output as FTEST input is Varied

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